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Evaluation of the Accuracy of RTDS VSC Models Determined from Frequency Scanning

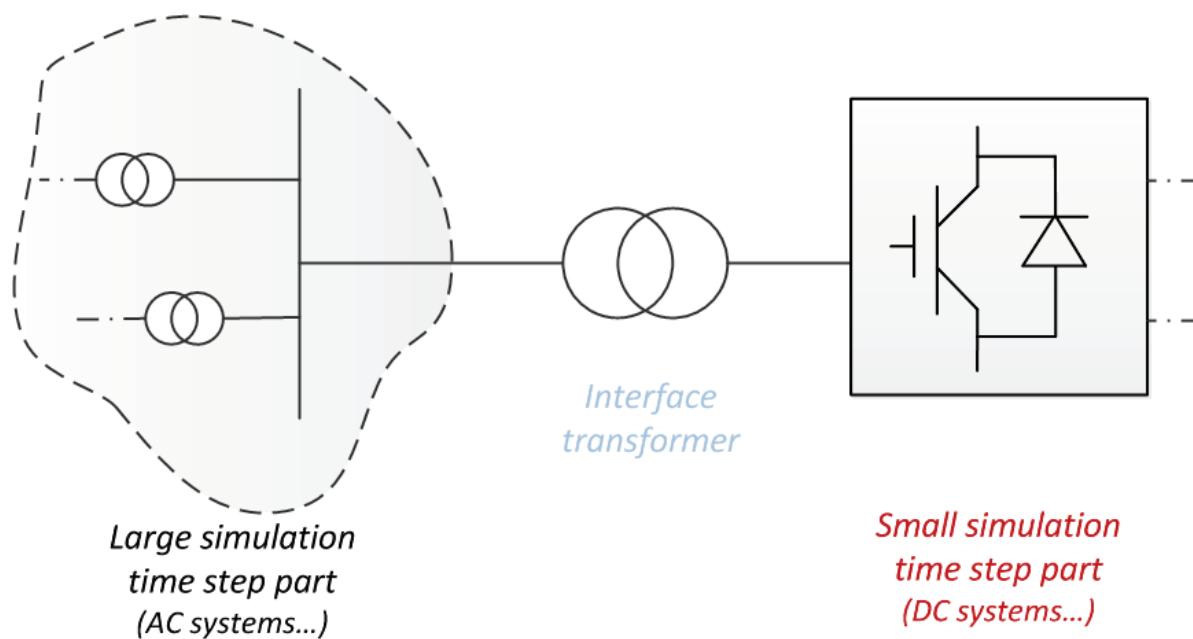
*Yi Qi and Ani Gole, University of Manitoba
Hui Ding and Yi Zhang, RTDS Technologies*



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VSC models (IGBT based) in RTDS

Small time step module



The 2-level converter is usually modeled using a small time step to improve accuracy...

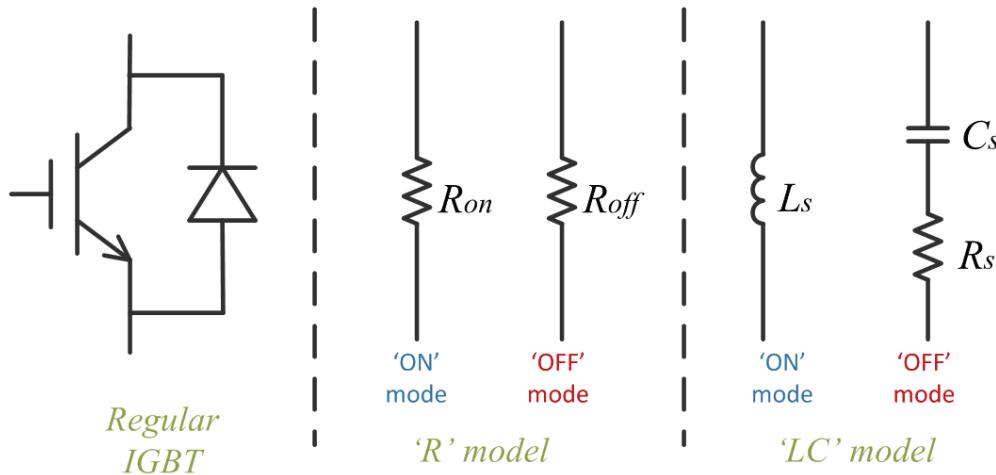


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Dommel Companion Circuit Representation for L and C

2-Level VSC models (IGBT based) in RTDS

Converter Model type ('R' and 'LC')



$$\frac{2L_s}{\Delta t_s} = \frac{\Delta t_s}{2C_s} + R_s$$

Dommel resistances are the same for 'ON' and 'OFF' states

LC Resonance?

'R' Type: 'ON' as a small resistor and 'OFF' as a large one

'LC' type: 'ON' as a small inductor and 'OFF' as a small capacitor in series with a resistor [1-2]

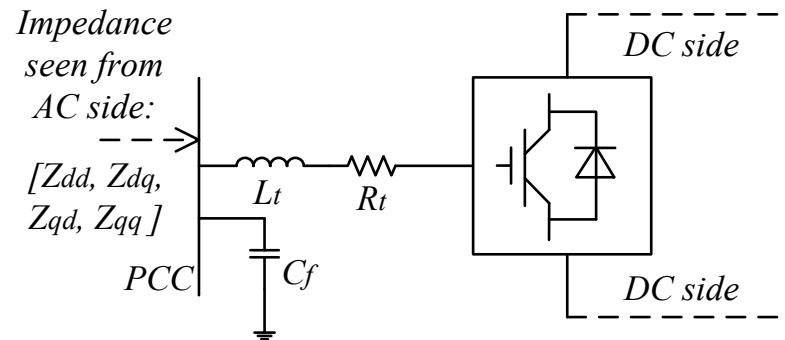


Converter Model Evaluation

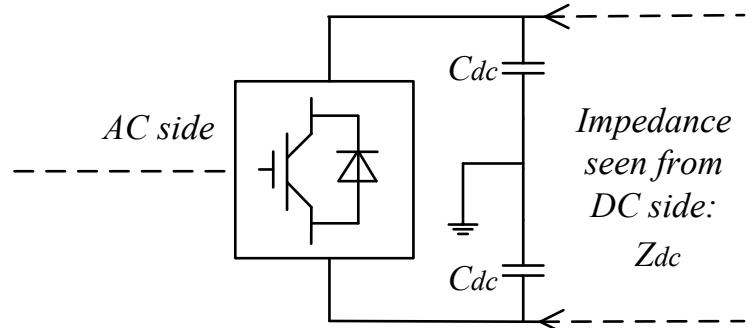
Methodology

Check the frequency response looking from the external terminals (ac and dc side) of the converter, which requires:

- A simple test system (for both 'R' and 'LC type')
- Frequency scanning method
- An analytical model for sanity check



a. impedance seen from AC side

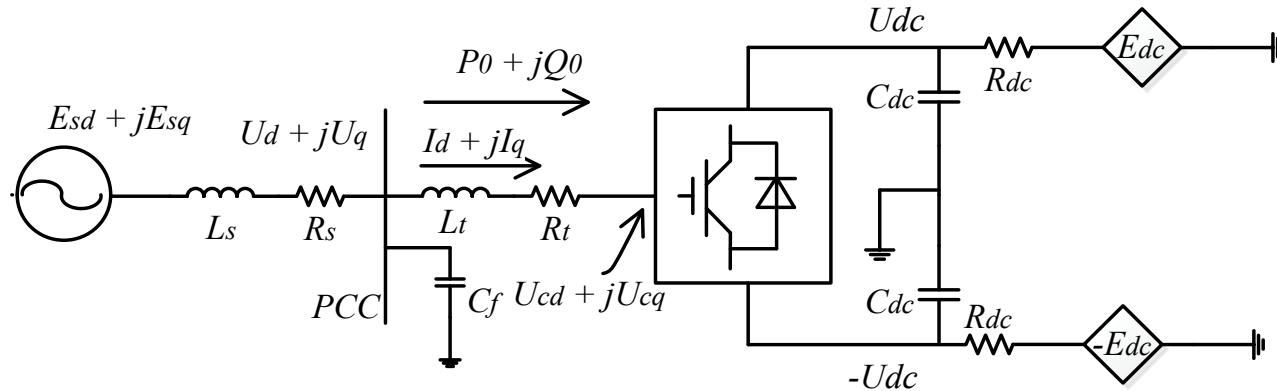


b. impedance seen from in AC side



Converter Model Evaluation

Case system



Case system: R-L ac system, ac filter, dc converter (either 'R' or 'LC'), dc capacitor and line resistance, a Phase Locked Loop as well as a decoupled controller.

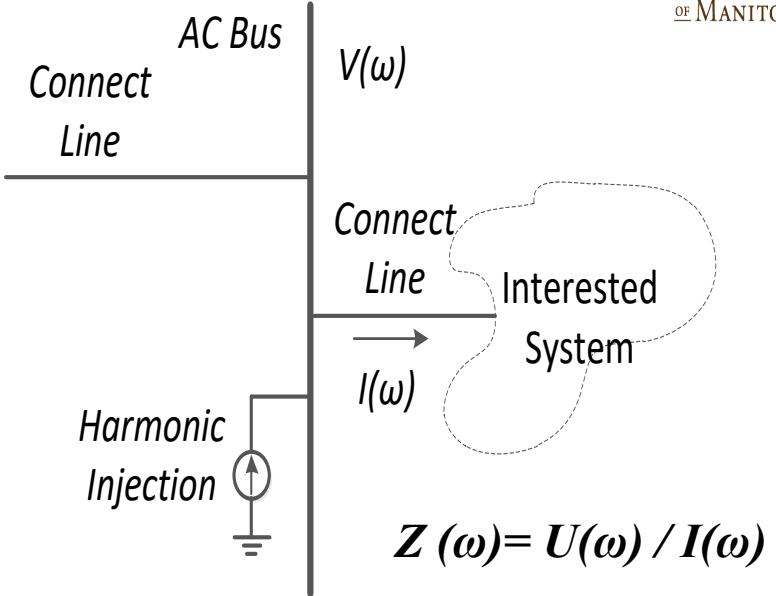
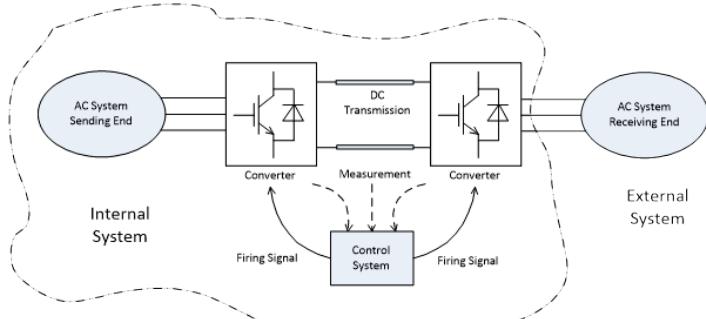
Simulation time step: 50us for large step and 2us for small step



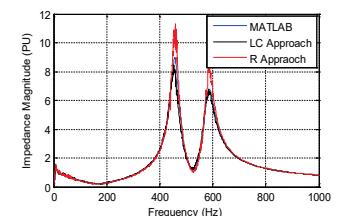
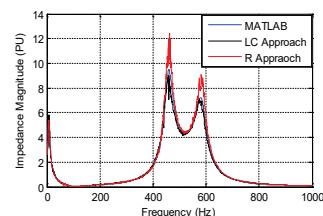
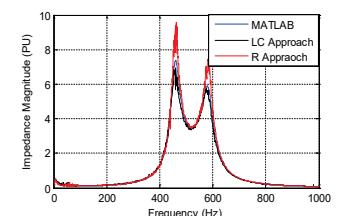
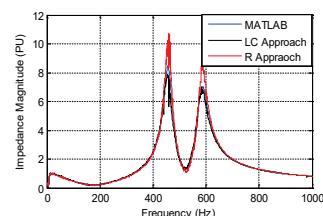
Network Stability using Impedance

Scan (Yi Qi, M.Das)

Ratio between
Voltage & current spectral
components under different
Frequencies. [1-2]



- ✓ Experimental approach to determine transfer function



Converter Model Evaluation

Frequency scanning

A. Multiple frequency injection [3]:

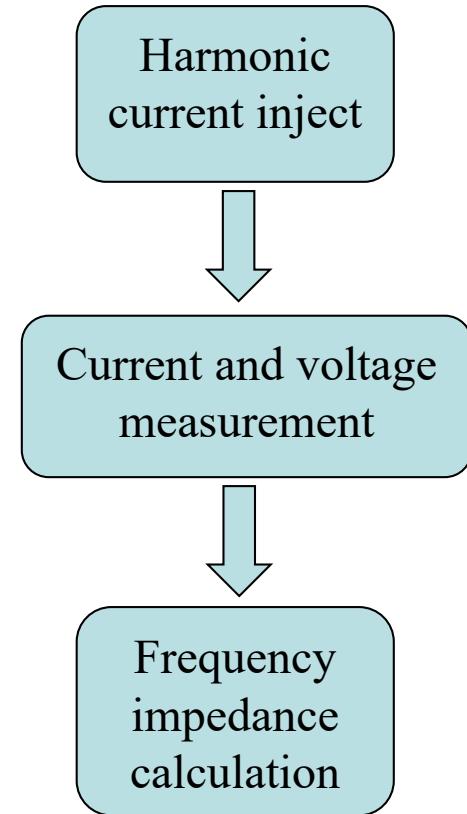
$$X_{inj}(t) = \sum_{f=f^{min}}^{f^{max}} X_{mag} \cos(2\pi ft + k_{inj}f^2)$$

B. Discrete Fourier Transform (DFT):

$$X_k = \frac{2}{N} \sum_{n=1}^N x_n e^{-j2\pi k \frac{n}{N}}$$

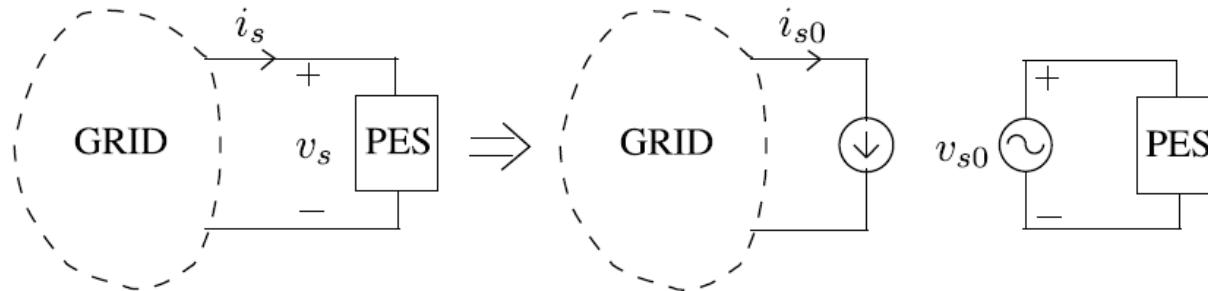
C. Impedance Result:

$$Y(f) = X(k\Delta f_o) / X_{inj} \quad (k\Delta f_o) = I(\omega) / V(\omega)$$

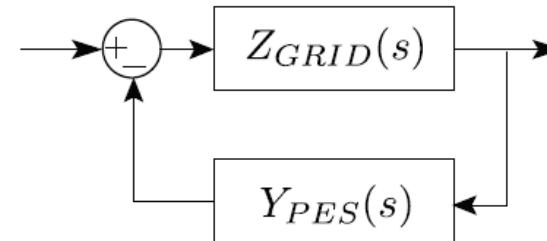


Grid Interaction Studies

Power Electronic System (PES) Connected to Grid:



Small Signal Model:



Closed Loop Transfer function:

$$\begin{aligned} T.F &= Z_{GRID}(s)(I + Y_{PES}(s)Z_{GRID}(s))^{-1} \\ &= (Z_{GRID}(s)^{-1} + Y_{PES}(s))^{-1} \end{aligned}$$

Converter Model Evaluation

Analytical model

The analytical model is built under

d-q domain [4]

considering the effect of *the ‘ON’ and ‘OFF’ resistance.*

AC side and DC side voltage-current relationship:

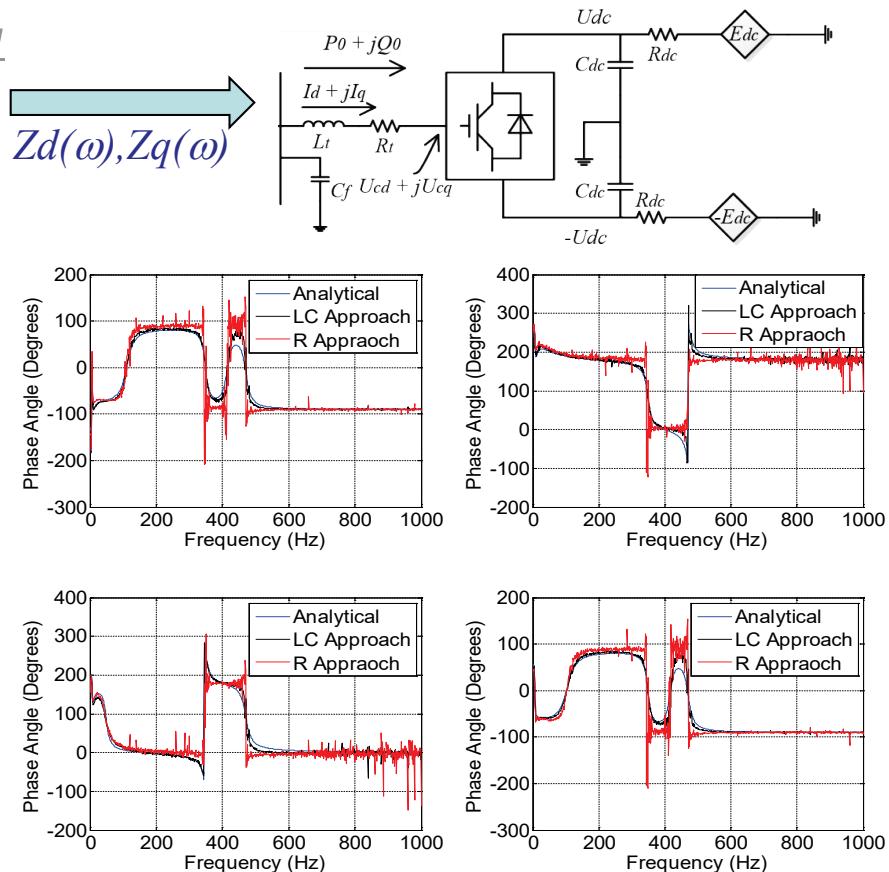
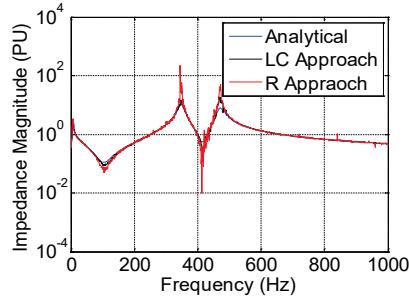
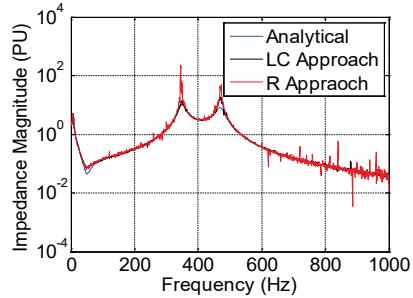
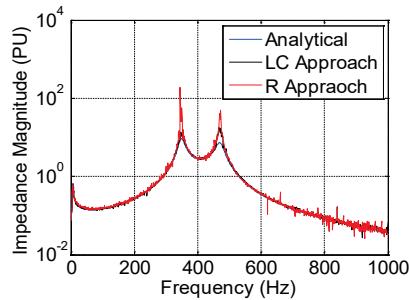
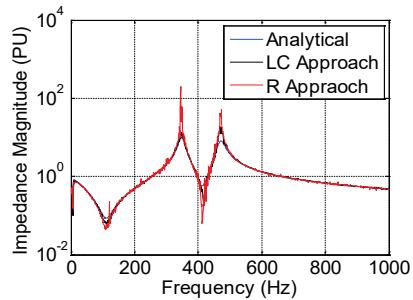
$$\begin{bmatrix} \Delta V_d(s) \\ \Delta V_q(s) \end{bmatrix} = \begin{bmatrix} \Delta V_d(s) \\ \Delta V_q(s) \end{bmatrix} = \begin{bmatrix} Z_{dd}(s) & Z_{dq}(s) \\ Z_{qd}(s) & Z_{qq}(s) \end{bmatrix} \begin{bmatrix} \Delta I_d(s) \\ \Delta I_q(s) \end{bmatrix}$$

$$\Delta V_{dc}(s) = Z_{dc}(s) \Delta I_{dc}(s)$$

This small signal result is dependent on the operating point.

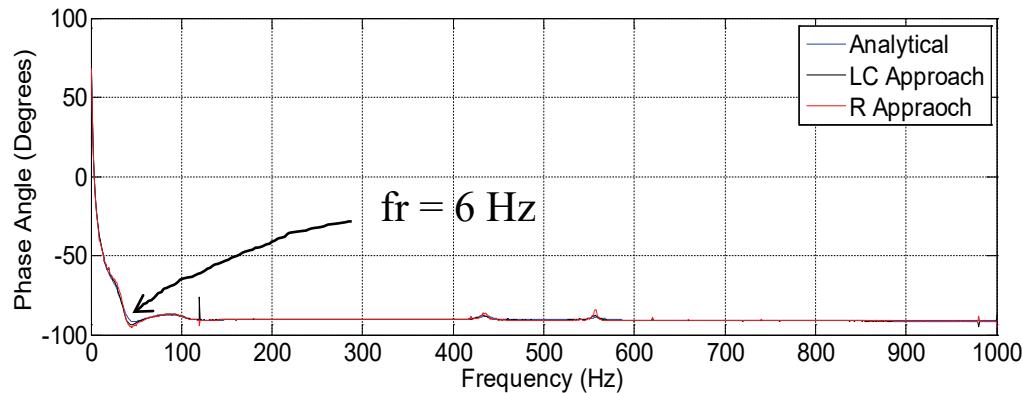
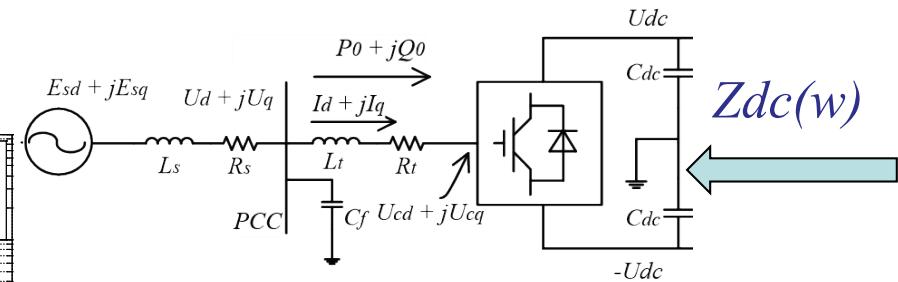
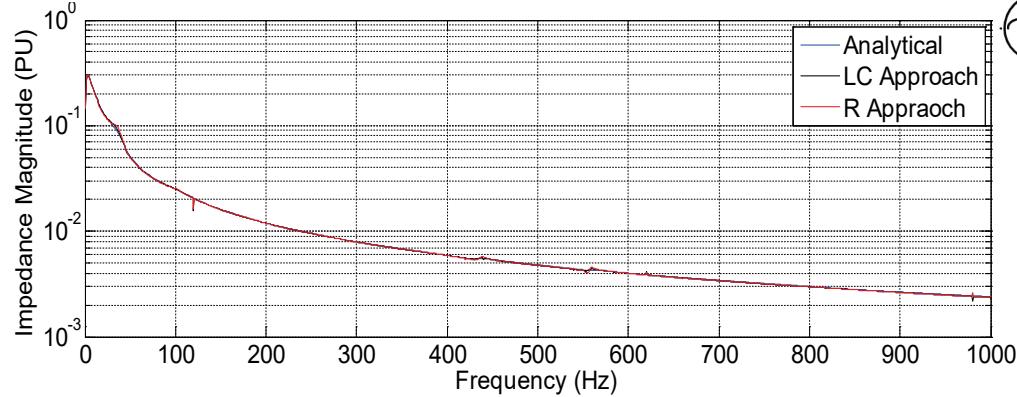


Comparison of Results, ac side: Operating point 1, $P_0 = 1.0$; $Q_0 = 0.0$



AC side impedance Z_{dd} Z_{dq} Z_{qd} Z_{qq} (OP 1), left: magnitude; right: phase angle

Comparison of Results, dc side: Operating point 1, $P_0 = 1.0; Q_0 = 0.0$



DC side impedance - Top: magnitude; Bottom: phase angle

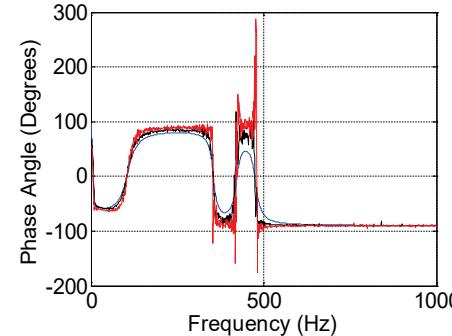
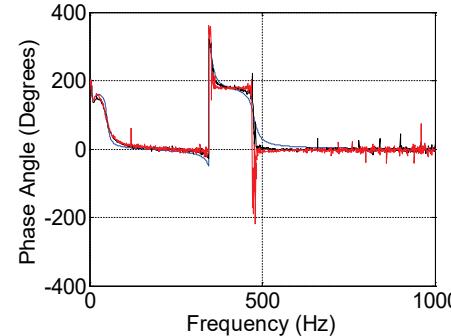
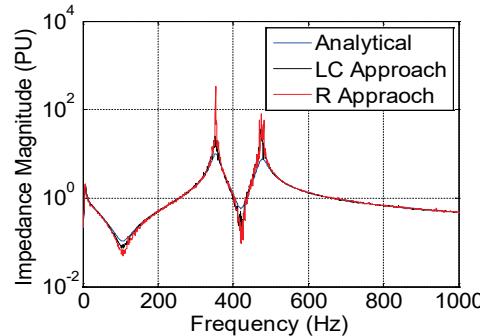
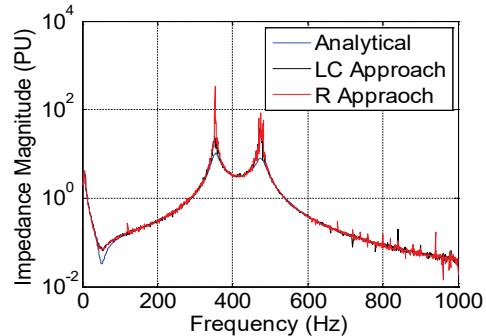
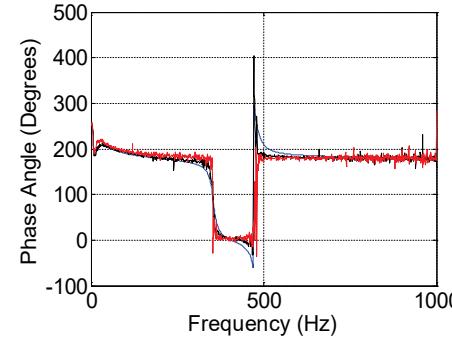
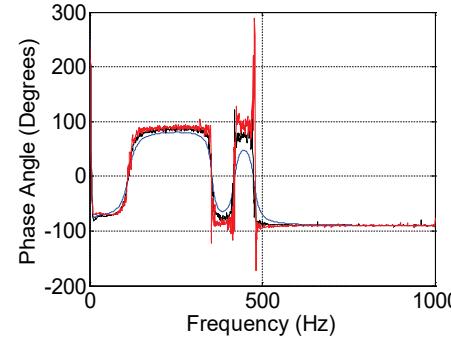
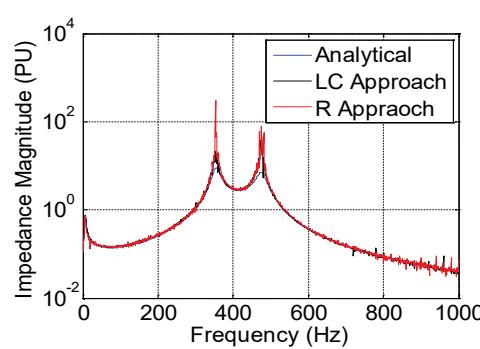
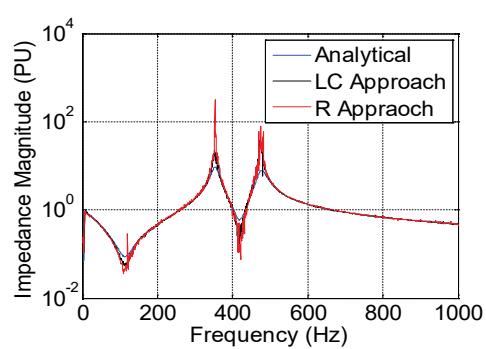
➤ Operating Point 1:
 $P_0 = 1.0; Q_0 = 0.0.$

➤ resonance: 6Hz



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Comparison of Results, ac side: Operating point 2,
 $P_0 = 0.6; Q_0 = 0.2.$

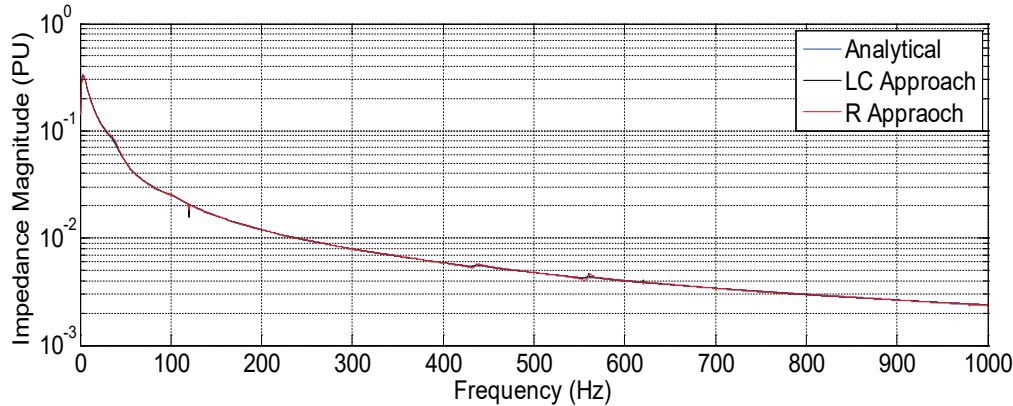


AC side impedance Z_{dd} Z_{dq} Z_{qd} Z_{qq} (OP 2), left: magnitude; right: phase angle

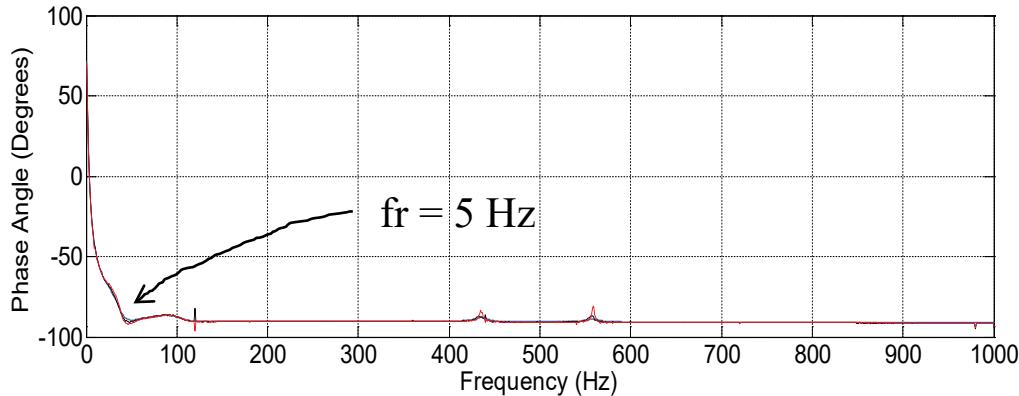


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Comparison of Results, ac side: Operating point 2, $P_0 = 0.6$; $Q_0 = 0.2$.



- *operating point 2:*
 $P_0 = 0.6; Q_0 = 0.2.$
- *resonance frequency:* 5Hz



DC side impedance - Top: magnitude; Bottom: phase angle



Concluding Remarks

- **The frequency response of the Converter is essentially unaffected by the LC representation**
- **This allows Power electronic subsystems to be effectively modelled with low computation overhead**
- **Although the tests were conducted on a 2-level converter, it is likely that MMCs and other PE models will also show good accuracy.**



References

- [1] RTDS Technologies, *Real time digital simulation for the power industry manual set*, Winnipeg, Canada, 2006.
- [2] Trevor Maguire and James Giesbrecht, "Small Time-step (<2uSec) VSC Model for the Real Time Digital Simulator", Proceeding of IPST 2005, Montreal Canada, June 2005, Paper No. IPST-168-25c.
- [3] Xiao Jiang and A. M. Gole, "A frequency scanning method for the identification of harmonic instabilities in HVDC systems," in *IEEE Transactions on Power Delivery*, vol. 10, no. 4, pp. 1875-1881, Oct 1995.
- [4] J. Z. Zhou, H. Ding, S. Fan, Y. Zhang and A. M. Gole, "Impact of Short-Circuit Ratio and Phase-Locked-Loop Parameters on the Small-Signal Behavior of a VSC-HVDC Converter," *IEEE Transactions on Power Delivery*, vol. 29, no. 5, pp. 2287-2296, Oct. 2014.
- [5] M. Mohaddes, A. M. Gole and S. Elez, "Steady state frequency response of STATCOM," in *IEEE Transactions on Power Delivery*, vol. 16, no. 1, pp. 18-23, Jan 2001.
- [6] T. Li, A. M. Gole and C. Zhao, "Harmonic Instability in MMC-HVDC Converters Resulting From Internal Dynamics," in *IEEE Transactions on Power Delivery*, vol. 31, no. 4, pp. 1738-1747, Aug. 2016.



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*Thank you very much!
May 2017*



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